

ASM-HEMT Model for GaN RF and Power Electronic Applications: Overview and Extraction

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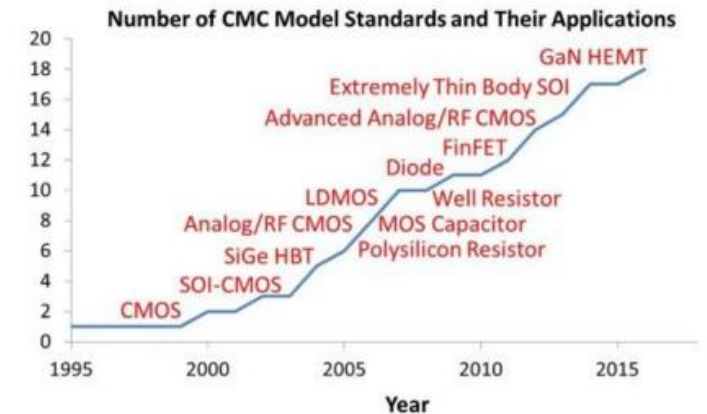
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GaN HEMT Model Standardization Effort in CMC

- Compact Model Coalition – an industry body that standardizes and promotes SPICE models for semiconductor devices as well as compiled modeling interface
- Dedicated workgroup for GaN HEMT model standardization launched in Year 2011
- Standardization process
 - Phase I completed: solicitation of models and presentation to CMC.
 - Phase II completed: shortlisted candidate models being examined against fundamental requirements and being fitted to measurement data for CMC evaluation
 - **Phase III is ongoing: evaluation on runtime, convergence, operability, etc.**
 - ASM-HEMT
 - MVSG
 - Phase IV: ballot for standardization (targeting at the end of 2016)

* Compact Model Coalition, https://www.si2.org/cmc_index.php

* Keith Green, http://www.eetimes.com/author.asp?section_id=36&doc_id=1324925



The CMC has been developing modeling standards for nearly two decades.

ASM-HEMT Model Overview

$$n_s = DV_{th} \left\{ \ln \left[\exp \left(\frac{E_f - E_0}{V_{th}} \right) + 1 \right] + \ln \left[\exp \left(\frac{E_f - E_1}{V_{th}} \right) + 1 \right] \right\}$$

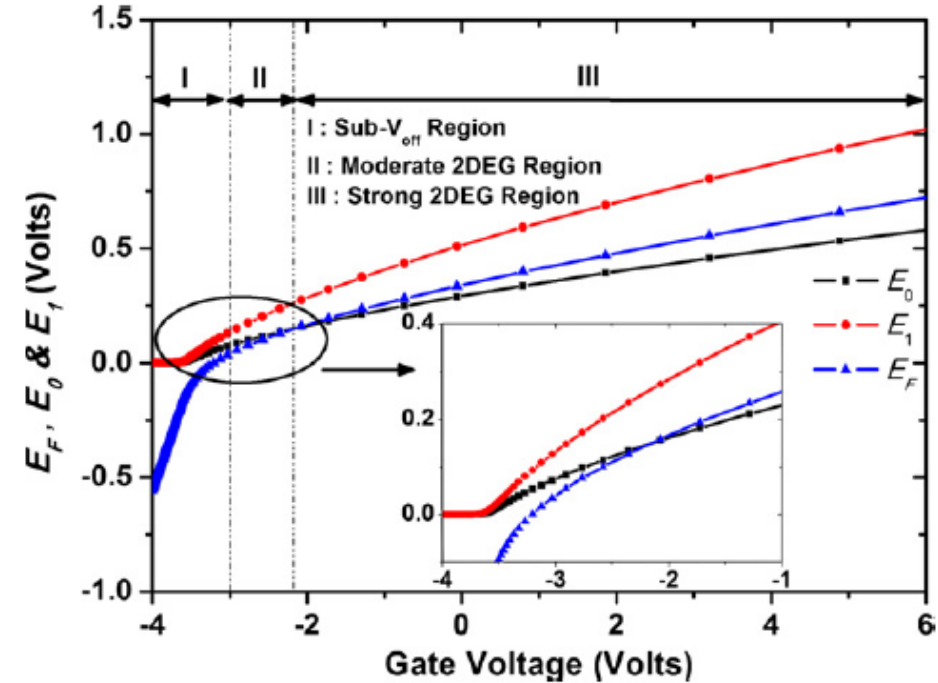
$$E_0 = \gamma_0 n_s^{2/3} \quad E_1 = \gamma_1 n_s^{2/3}$$

$$n_s = \frac{\epsilon}{qd} (V_{go} - E_f - V_x)$$

Transcendental

$$E_{f,unified} = V_{go} - \frac{2V_{th} \ln \left(1 + e^{\frac{V_{go}}{2V_{th}}} \right)}{\frac{1}{H(V_{go}, p)} + \frac{C_g}{qD} e^{-\frac{V_{go}}{2V_{th}}}}$$

$$\psi = E_f + V_x$$



ASM HEMT Model Overview

$$I_d(x) = \mu W N_f Q_{ch} \frac{d\psi(x)}{dx} + \mu W N_f V_{th} \frac{dQ_{ch}}{dx} \quad Q_{ch} = -C_g(V_{go} - \psi(x))$$

$$I_d = \frac{W}{L} \mu C_g N_f \left[V_{go} - \left(\frac{\psi_s + \psi_d}{2} \right) + V_{th} \right] (\psi_d - \psi_s)$$

Intrinsic Charges

$$Q_{gi} = - \int_0^L W N_f C_g (V_{go} - \psi(x)) dx$$

$$Q_{di} = \int_0^L (x/L) W N_f C_g (V_{go} - \psi(x)) dx$$

$$Q_{si} = -Q_{gi} - Q_{di}$$

Incorporating Realistic device effects

$$I_d = \frac{\mu_{eff}}{\sqrt{1 + \theta_{sat}^2 \psi_{ds}^2}} \frac{W}{L} C_g N_f \left[V_{go} - \left(\frac{\psi_s + \psi_d}{2} \right) + V_{th} \right] \psi_{ds} (1 + \lambda V_{ds})$$

List of Parameters I

Core Model Parameters

```
parameter real VOFF = -2.0           from [-100.0:5];           //Cut-off voltage
parameter real U0 = 170.0e-3        from [0.0:inf);           //Low field mobility
parameter real UA = 0.0e-9          from [0.0:inf);           //Mobility Degradation coefficient first order
parameter real UB = 0.0e-18        from [0.0:inf);           //Mobility Degradation coefficient second order
parameter real VSAT = 1.9e5         from [1.0e3:inf);         //Saturation Velocity
parameter real DELTA = 2.0          from [2.0:inf);           //Exponent for Vdeff
parameter real AT = 0.0             from (-inf:inf);          //Temperature Dependence for saturation velocity
parameter real UTE = -0.5           from [-10.0:0];           //Temperature dependence of mobility
parameter real LAMBDA = 0.0         from [0.0:inf);           //Channel Length Modulation Coefficient
parameter real ETA0 = 1.0e-9        from [0.0:inf);           //DIBL Parameter
parameter real VDSCALE = 5.0        from (0.0:inf);           //DIBL Scaling VDS
parameter real KT1 = 0.0e-3         from (-inf:inf);          //Temperature Dependence for Voff
parameter real THESAT = 1.0         from [1.0:inf);           //Velocity Saturation Parameter
parameter real NFACTOR = 0.5        from [0.0:inf);           //Sub-VOFF Slope parameters
parameter real CDSCD = 1.0e-3       from [0.0:inf);           //Sub-VOFF Slope Change due to Drain Voltage
parameter real IMIN = 1.0e-15       from [0.0:inf);           //Minimum Drain Current
```

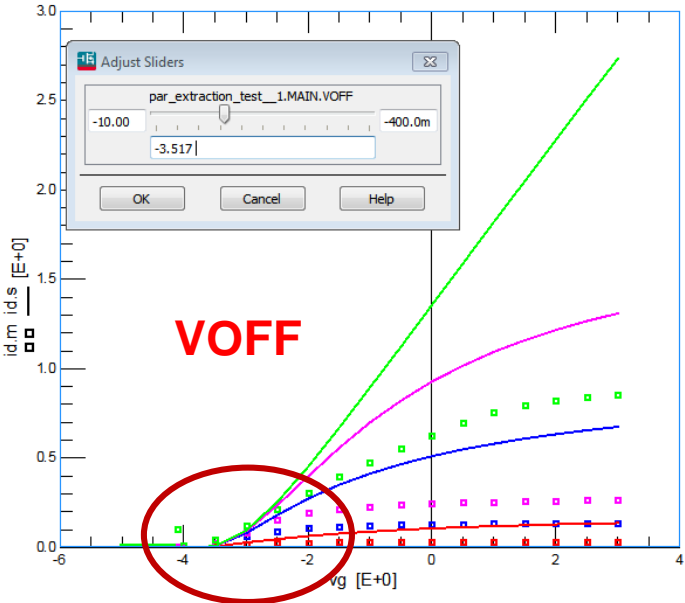
List of Parameters II

Access Region and Temperature Parameters

```
parameter integer RDSMOD = 1           from [0:1];           //Switch for external source and drain resistances
parameter real VSATACCS = 50.0e3       from [0.0:inf);      //Saturation Velocity for access region: Source Side
parameter real NSOACCS = 5.0e17        from [1.0e5:inf);    //2-DEG Charge Density in per square meter in Source access region
parameter real NSOACCD = 5.0e17       from [1.0e5:inf);    //2-DEG Charge Density in per square meter in Drain access region
parameter real KOACCS = 0.0            from [0.0:inf);      //Vg dependence parameter of source side access region 2-DEG charge density
parameter real KOACCD = 0.0           from [0.0:inf);      //Vg dependence parameter of drain side access region 2-DEG charge density
parameter real UOACCS = 155e-3         from [0.0:inf);      //Access region mobility source-side
parameter real UOACCD = 155e-3        from [0.0:inf);      //Access region mobility drain-side
parameter real MEXPACCS = 2.0          from [0.0:inf);      //Exponent for access region resistance model
parameter real MEXPACCD = 2.0         from [0.0:inf);      //Exponent for access region resistance model
parameter real LSG = 1.0e-6           from [0.0:inf);      //Length of Source-Gate Access Region
parameter real LDG = 1.0e-6           from [0.0:inf);      //Length of Drain-Gate Access Region or Length of drain side access region
parameter real RSC = 1.0e-4           from [0.0:inf);      //Source Contact Resistance
parameter real RDC = 1.0e-4           from [0.0:inf);      //Drain Contact Resistance
parameter real KNS0 = 0.0             from [0.0:inf);      //Temperature Dependence for 2-DEG charge density at access region
parameter real ATS = 0.0              from (-inf:inf);     //Temperature Dependence for saturation velocity at access region
parameter real UTES = 0.0             from (-inf:inf);     //Temperature dependence of mobility at access region: Source Side
parameter real UTED = 0.0            from (-inf:inf);     //Temperature dependence of mobility at access region: Drain Side
parameter real KRSC = 0.0            from [0.0:inf);      //Temperature dependence of Source Contact Resistance
parameter real KRDC = 0.0            from [0.0:inf);      //Temperature dependence of Drain Contact Resistance
parameter integer SHMOD = 1 ;          //Switch to turn on and off self-heating model
parameter real RTH0 = 5.0             from [0.0:inf);      //Thermal Resistance
parameter real CTH0 = 1.0e-9         from [0.0:inf);      //Thermal Capacitance
```

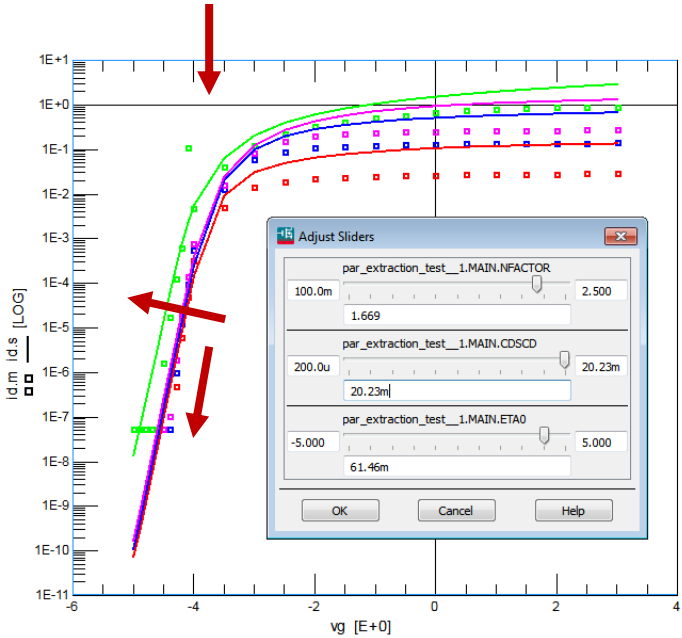
Parameter Extraction I

Id-Vg Linear and Log Scale (Linear Vd Condition)

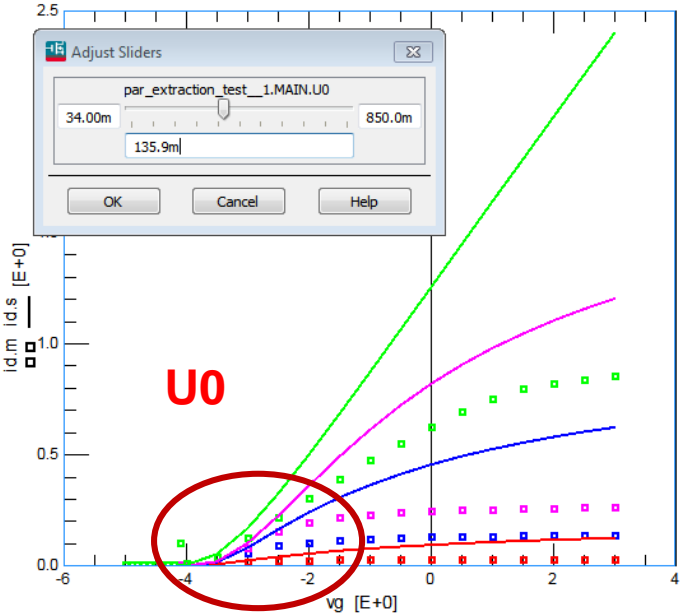


Id-Vgs (Linear)

NFACTOR
ETA0
VDSCALE
CDSCD



Id-Vgs (Log)



Id-Vgs (Linear)

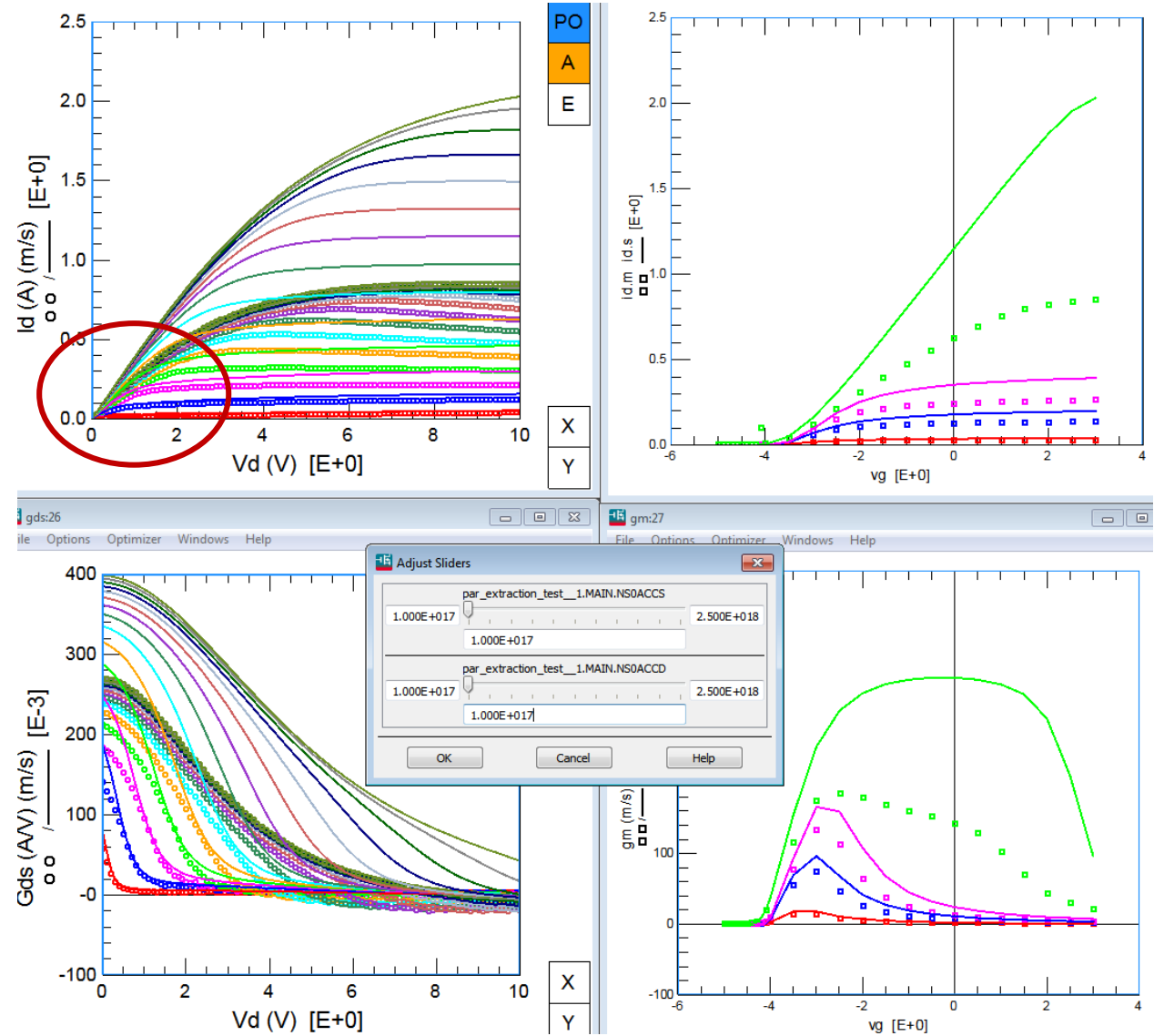
Parameter Extraction II

Id-Vgs, gm-Vgs and Id-Vds, gds-Vds

Adjust Ron using

NS0ACCS

NS0ACCD



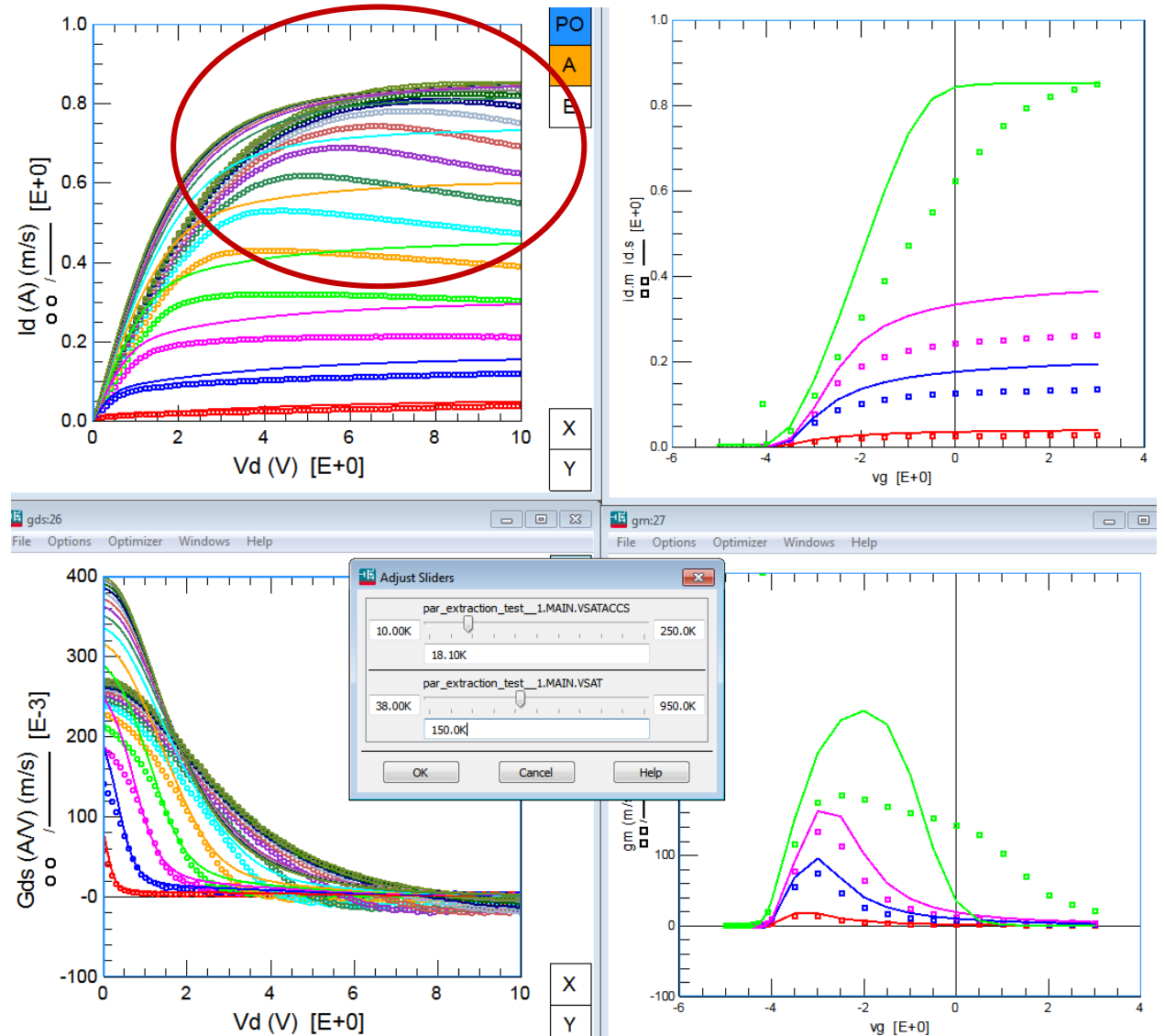
Parameter Extraction III

I_d - V_{gs} , g_m - V_{gs} and I_d - V_{ds} , g_{ds} - V_{ds}

Fit Saturation Current by adjusting

VSATACCS

VSAT should remain more or less the same since it is a material dependent parameter and should not vary much for different GaN Technologies



Parameter Extraction IV

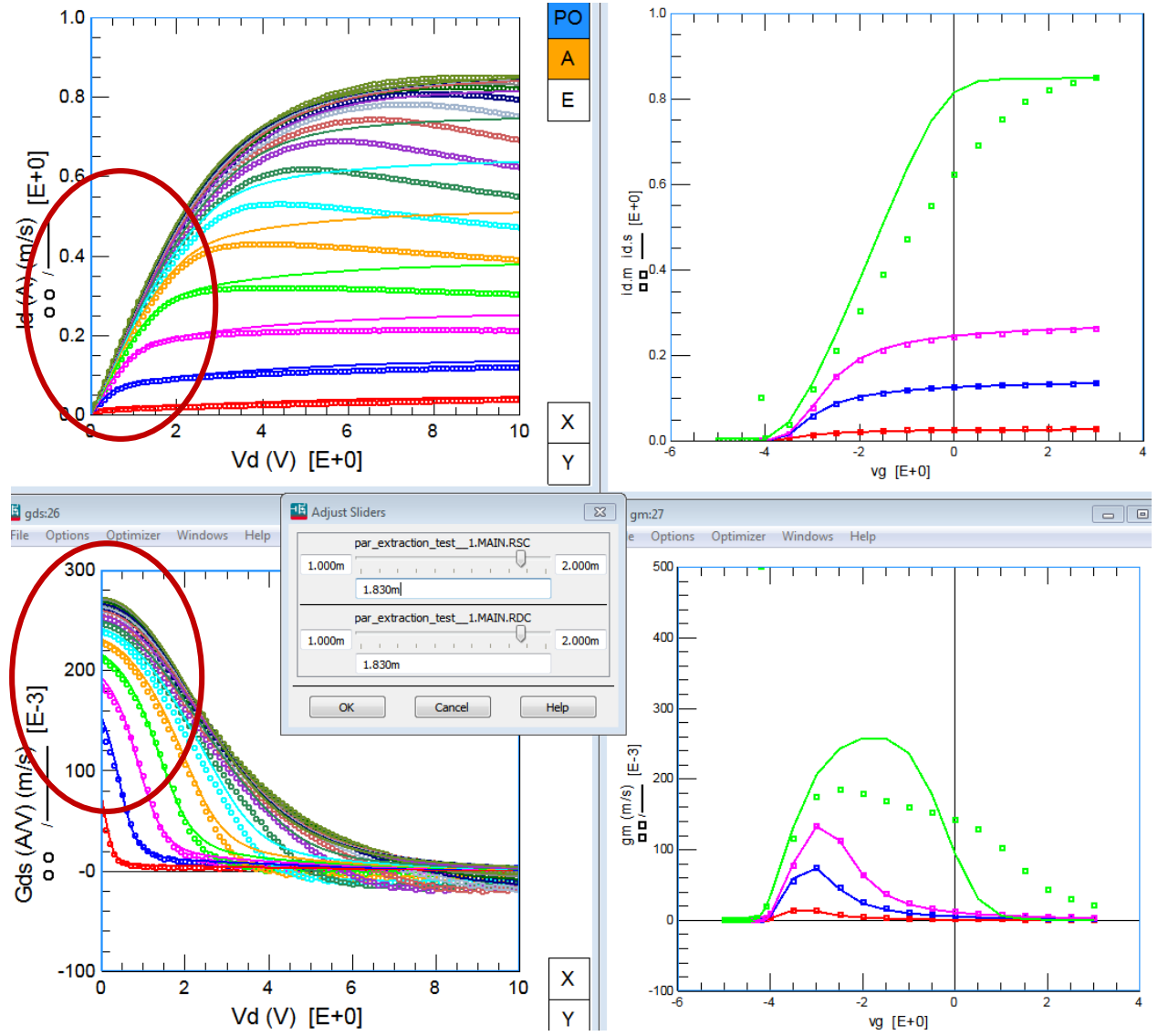
Id-Vgs, gm-Vgs and Id-Vds, gds-Vds

Fine tune Contact Resistance parameters

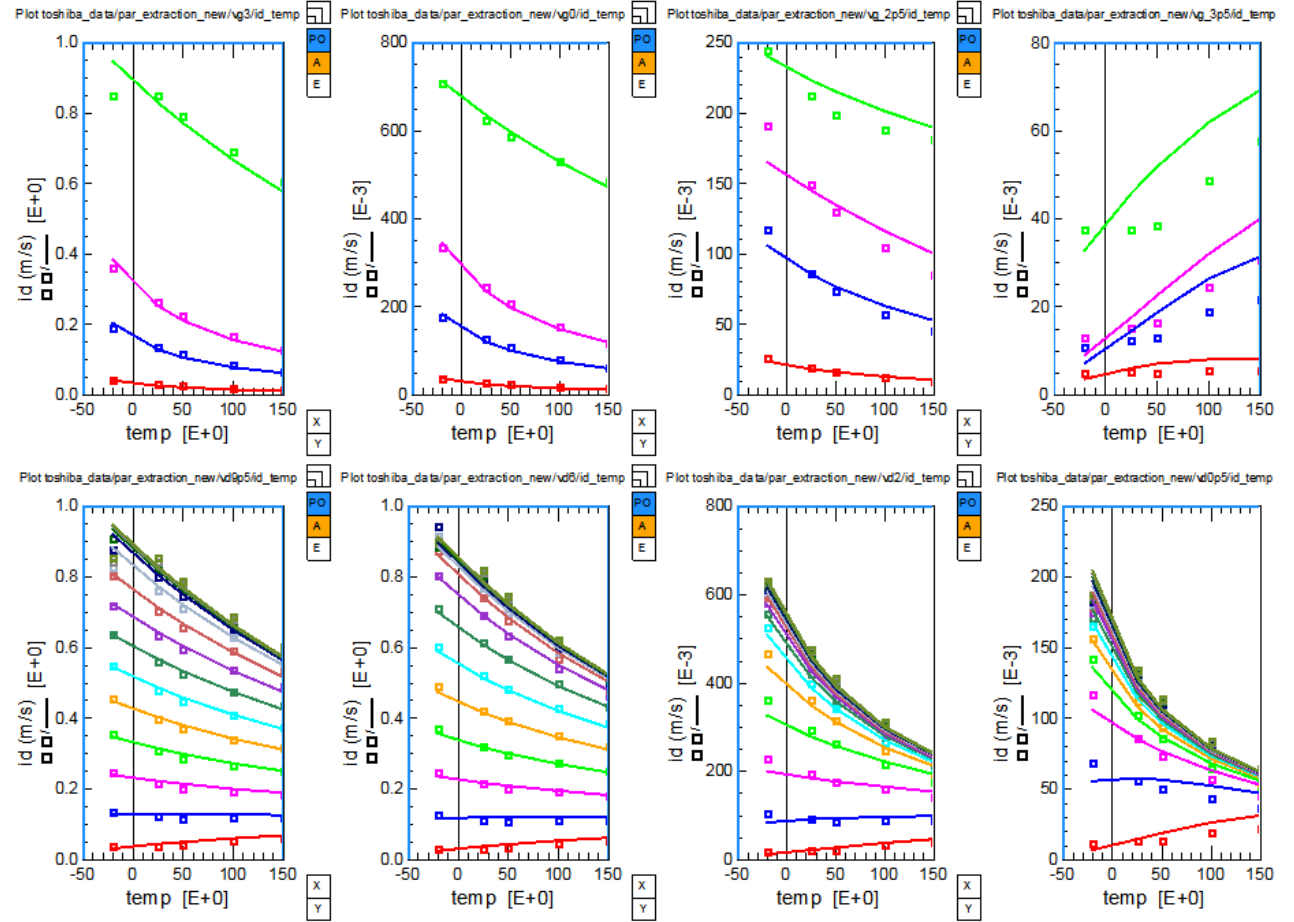
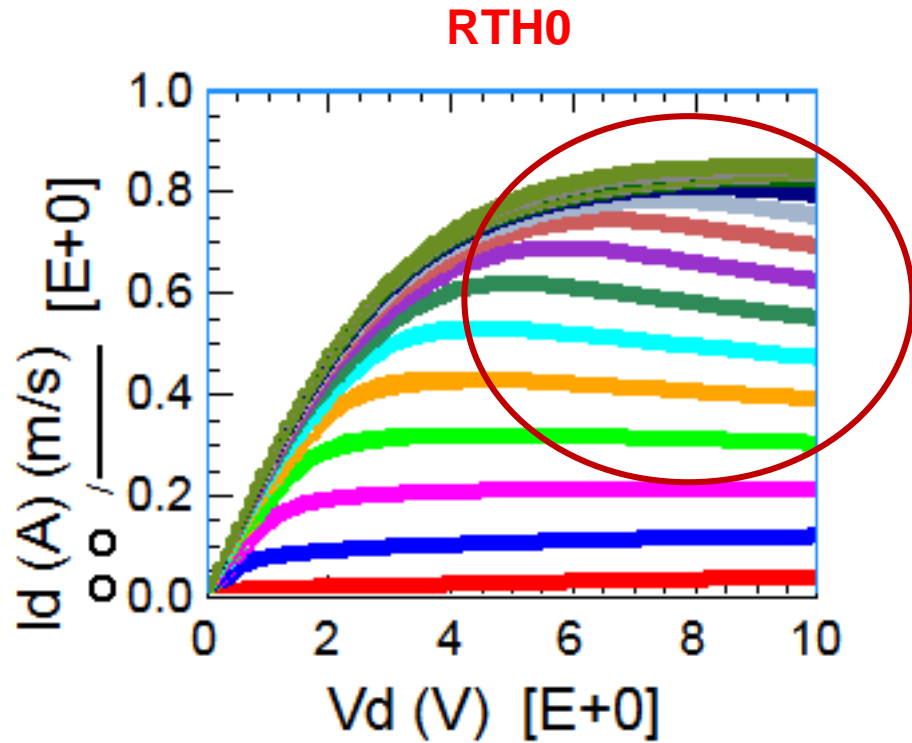
RDC

RSC

To further fit the Ron



Self Heating and Temperature Related Parameters



KT1

AT

ATS

KRSC

KRDC

UTE

Field Plate Capacitance Model

Terminal Capacitances

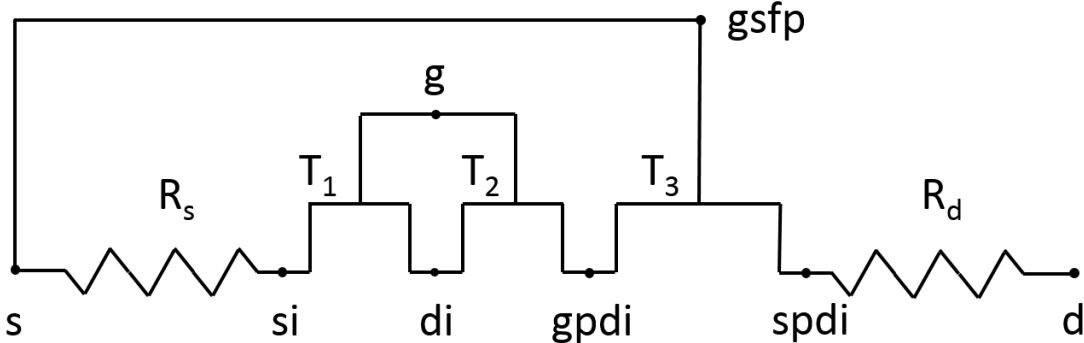
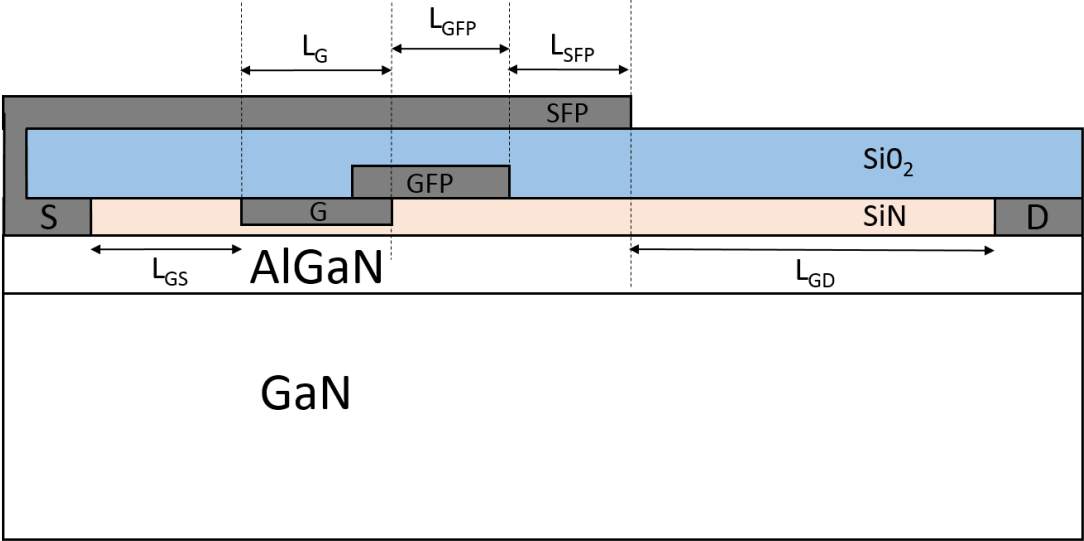
$$C_{iss} = C_{gs} + C_{gd}$$

$$C_{rss} = C_{gd}$$

$$C_{oss} = C_{ds} + C_{gd}$$

FP Modeled as an intrinsic HEMT

Each HEMT governed by ψ Calc.



Field Plate MOD Parameter Extraction I

Ciss – Vgs

$-150V < V_{gs} < 0V$

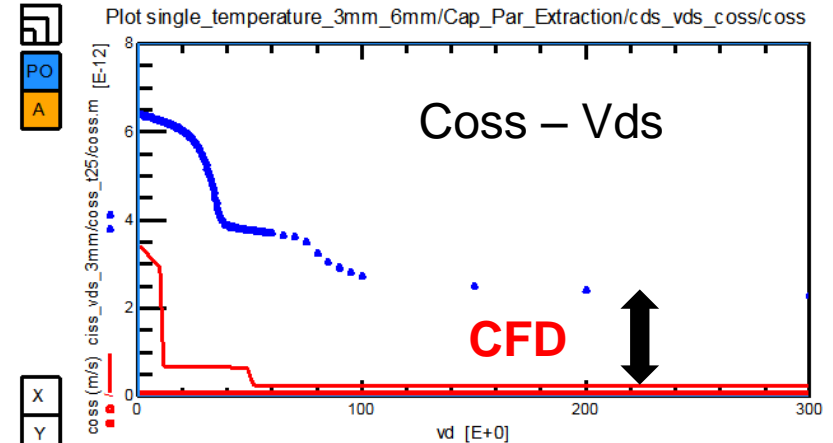
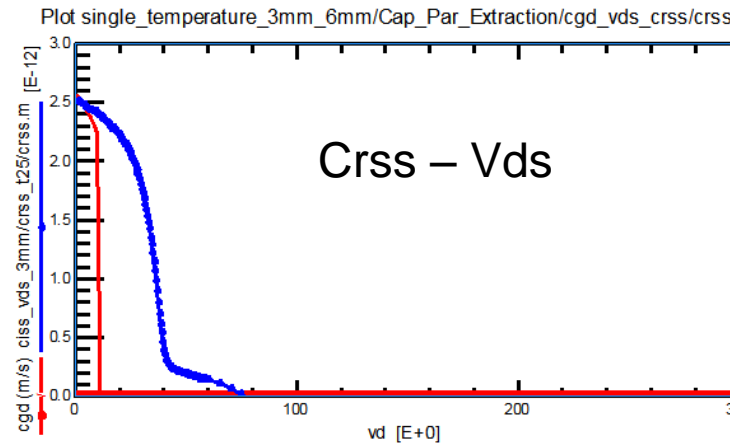
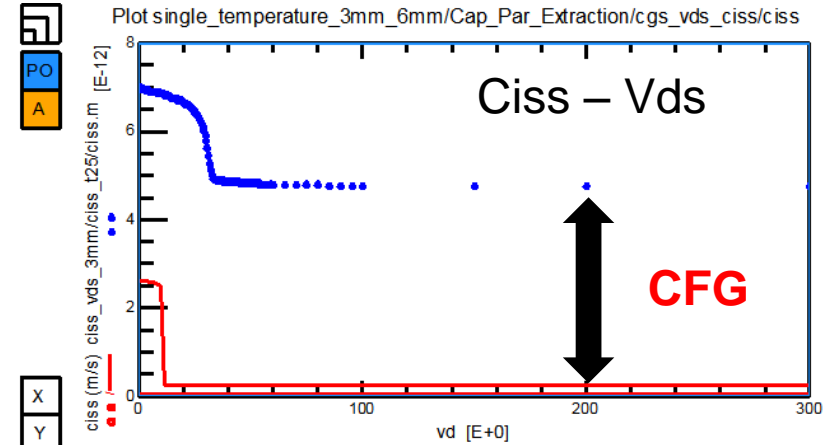
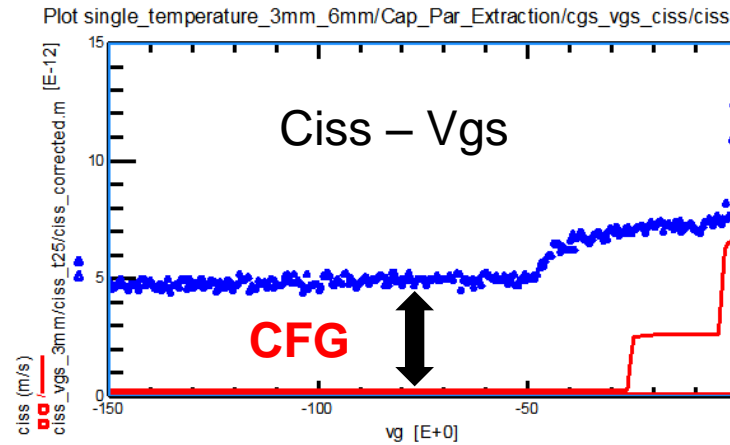
$V_{ds} = 0$

Crss, Ciss, Coss versus Vds

$0V < V_{ds} < 300V$

$V_{gs} = -15V$

GFPMOD = SFPMOD = 1

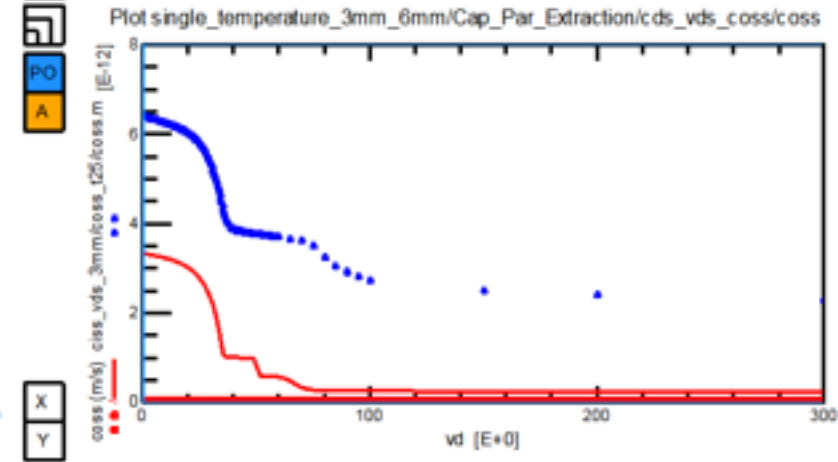
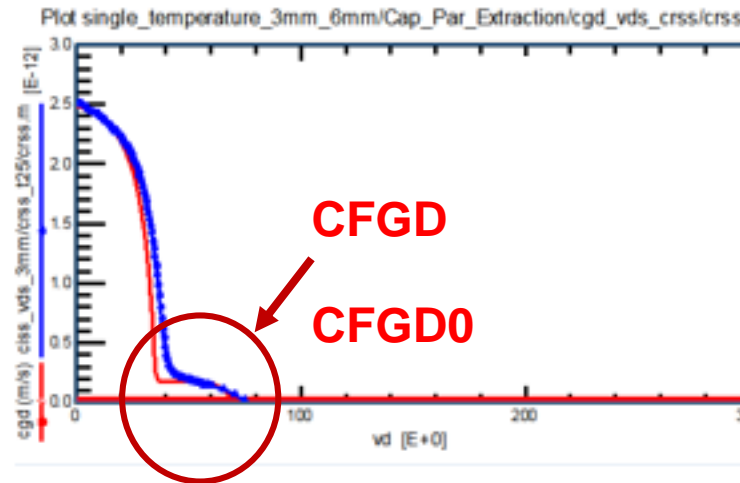
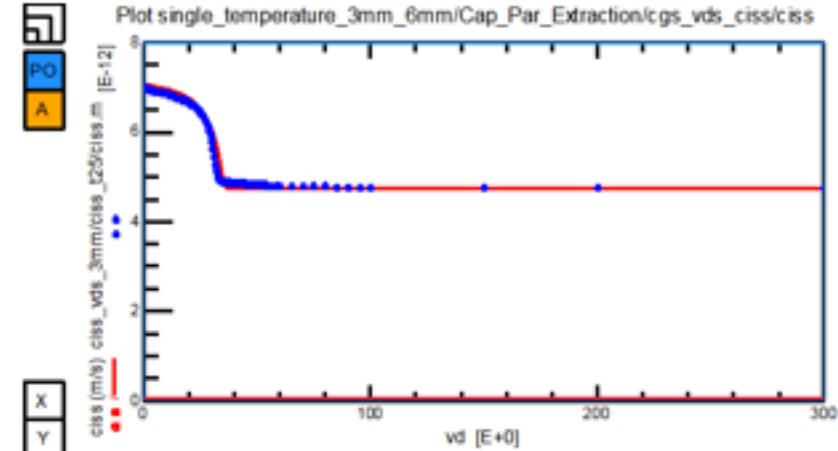
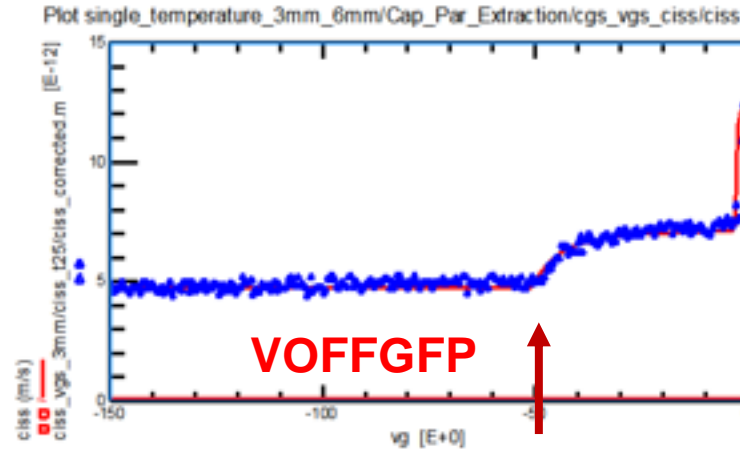


Field Plate MOD Parameter Extraction II

VOFFGFP – sets the rise of the hump

ADOS – Smoothens the hump

CFGD } Gate FP - Drain
CFGD0 } Fringing capacitance



Field Plate MOD Parameter Extraction III

Optimize the following

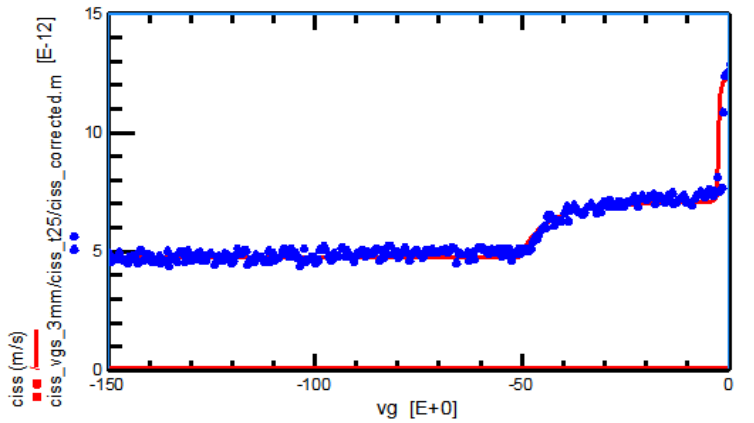
AJ

CJ0

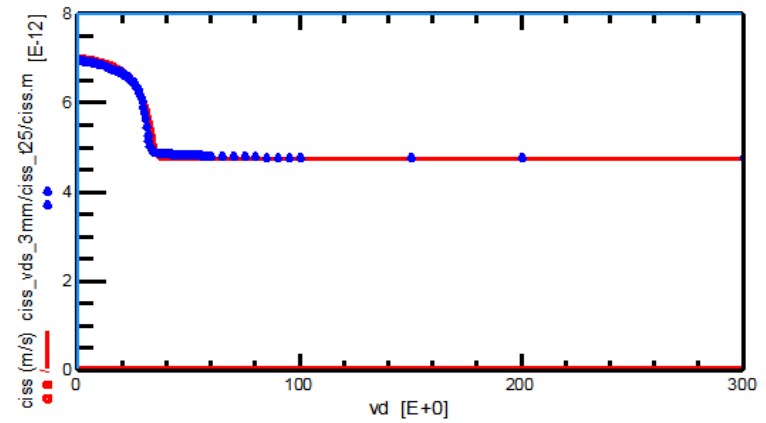
MZ

VBI

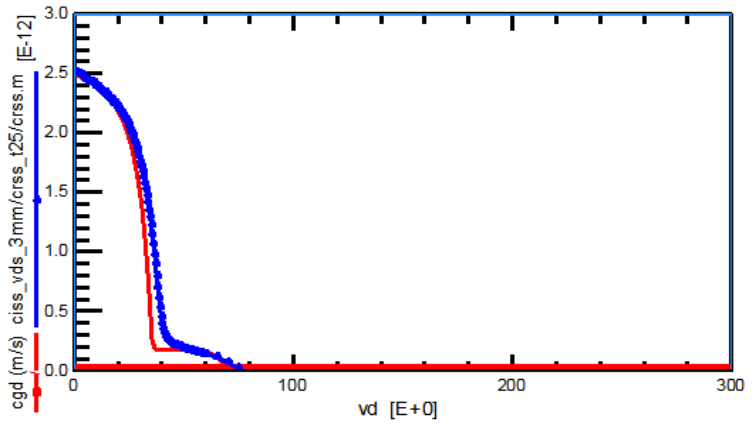
Plot single_temperature_3mm_6mm/Cap_Par_Extraction/cgs_vgs_ciss/ciss



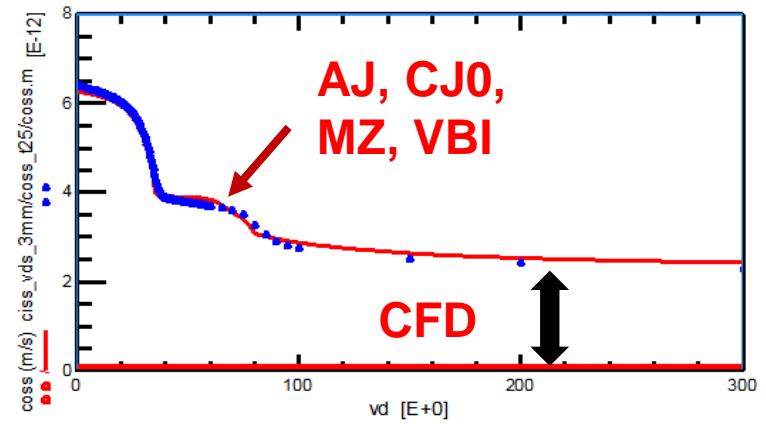
Plot single_temperature_3mm_6mm/Cap_Par_Extraction/cgs_vds_ciss/ciss



Plot single_temperature_3mm_6mm/Cap_Par_Extraction/cgd_vds_crss/crss



Plot single_temperature_3mm_6mm/Cap_Par_Extraction/cds_vds_coss/coss



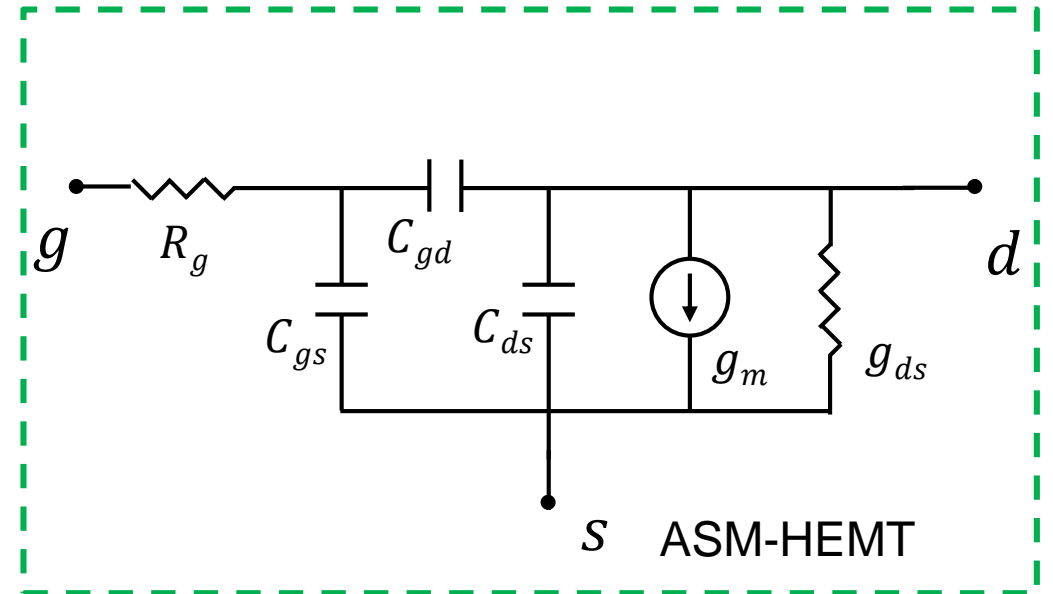
ASM HEMT - Small Signal Representation

$$Y_{11} = \frac{j\omega(C_{gs} + C_{gd})}{1 + \omega^2(C_{gs} + C_{gd})^2 R_g^2} + \frac{\omega^2(C_{gs} + C_{gd})^2 R_g}{1 + \omega^2(C_{gs} + C_{gd})^2 R_g^2}$$

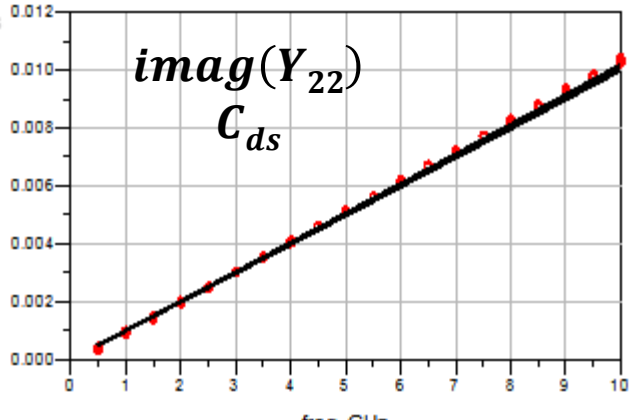
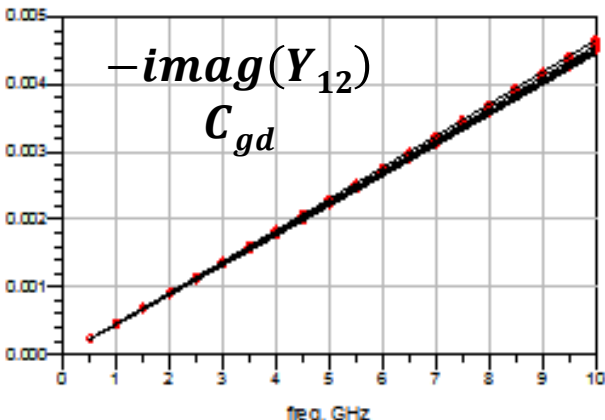
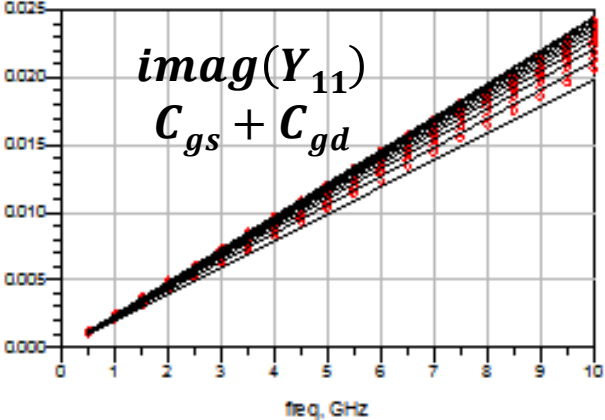
$$Y_{12} = -j\omega C_{gd} - \omega^2 C_{gd}(C_{gs} + C_{gd})R_g$$

$$Y_{21} = [g_m - \omega^2 C_{gd}(C_{gs} + C_{gd})R_g] - j\omega\{C_{gd} + g_m(C_{gs} + C_{gd})R_g\}$$

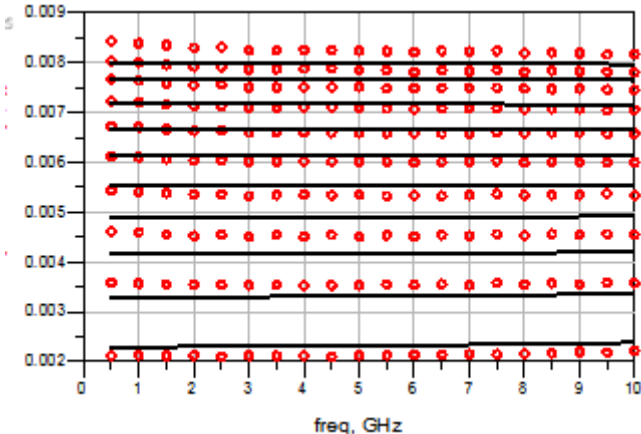
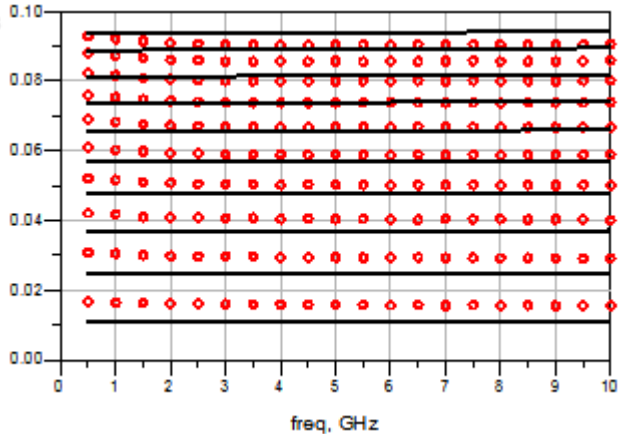
$$Y_{22} = g_{ds} + j\omega\{C_{ds} + C_{gd}(1 + g_m R_g)\}$$



De-embedded Y-Parameters (0.5-10 GHz)

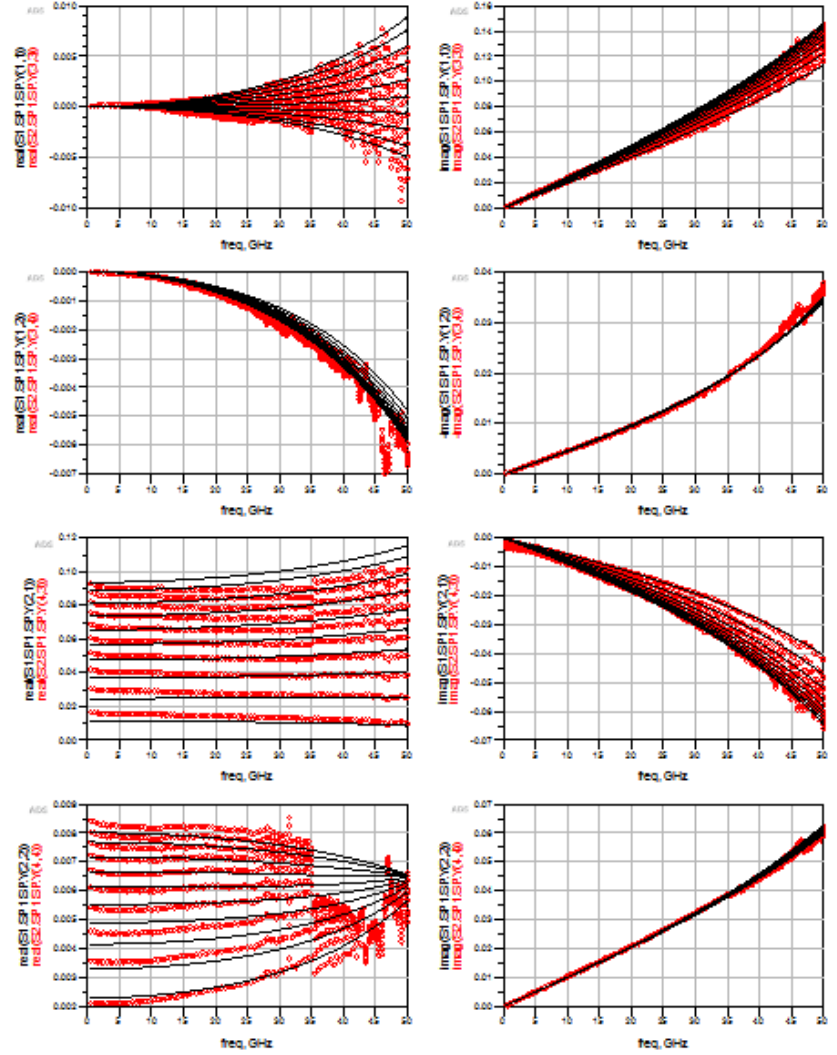
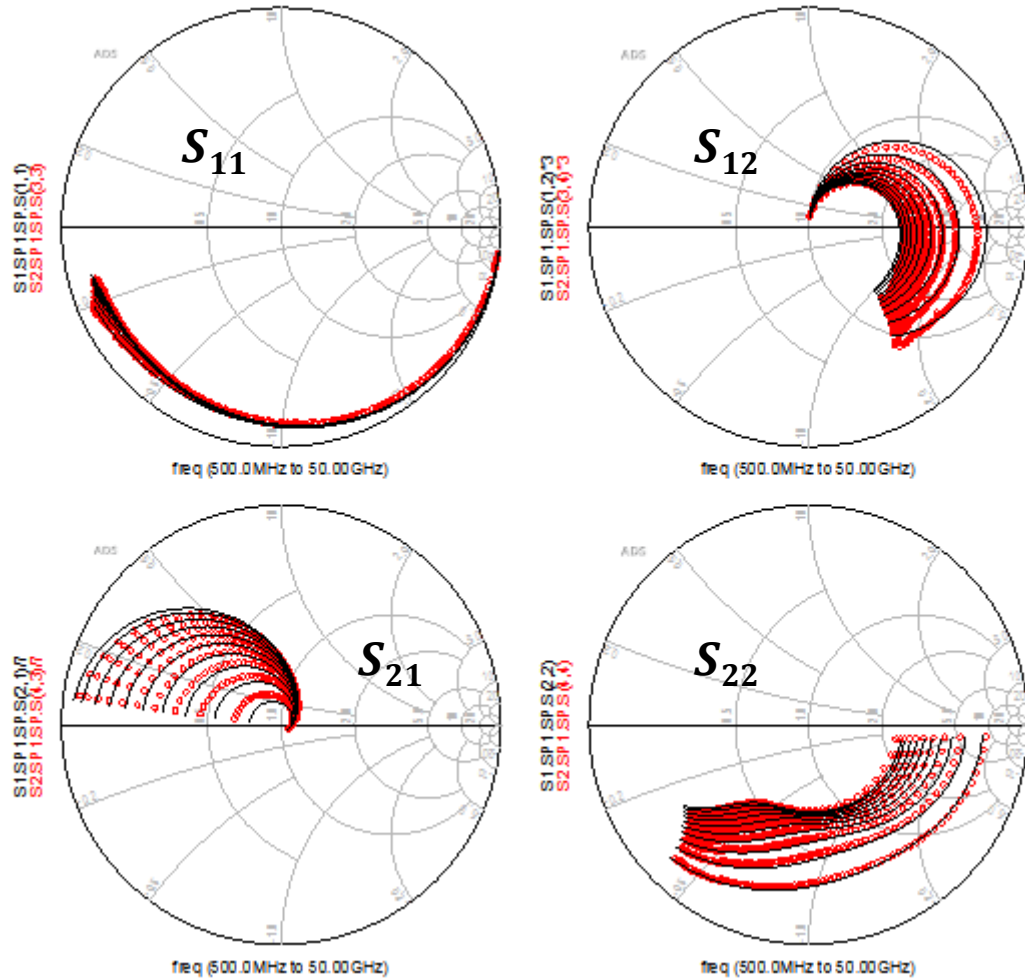


$real(Y_{21})$
 g_m



$real(Y_{22})$
 g_{ds}

Broadband S-Parameters (0.5-50 GHz)



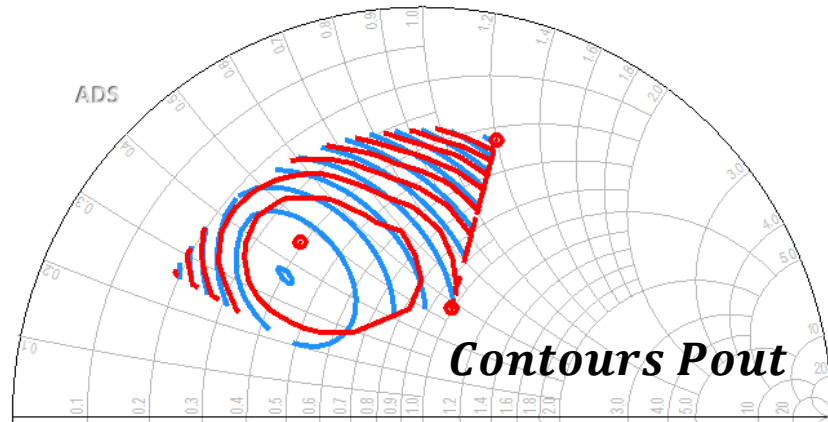
Y_{11}

Y_{12}

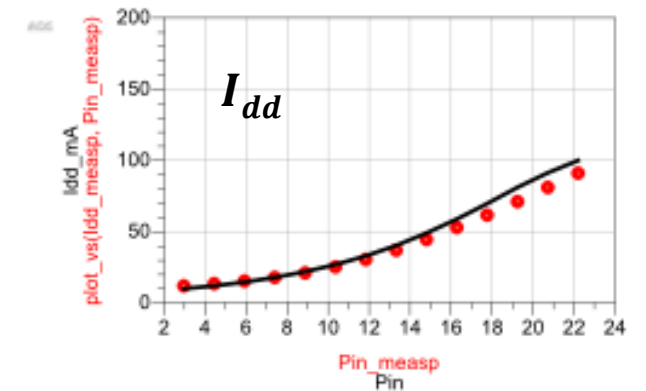
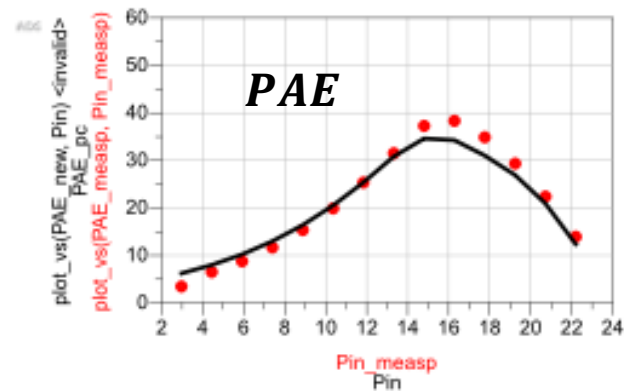
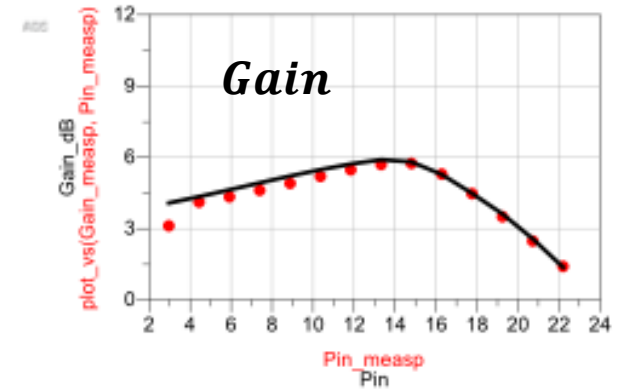
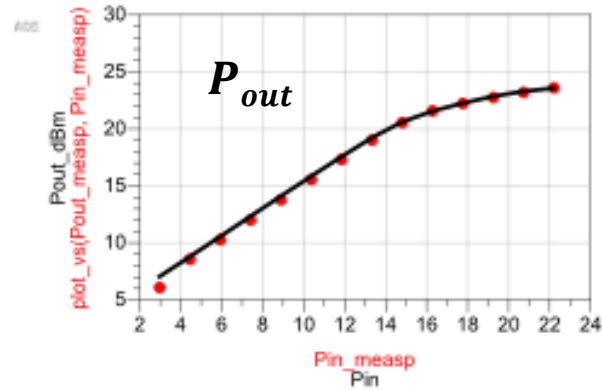
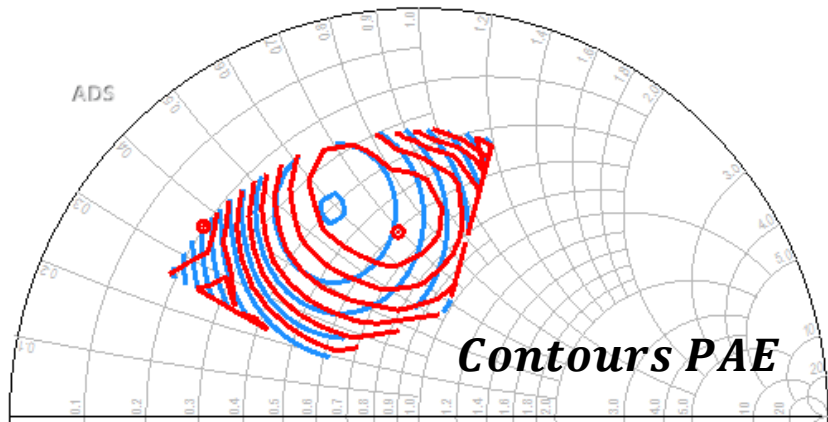
Y_{21}

Y_{22}

Load-Pull and Harmonic Balance Simulations



— Measured — Model



Summary

- As one of the candidates for CMC GaN HEMT, ASM-HEMT is a physical model base on surface potential analytical calculation
- Model extraction procedure
 - DC parameter extraction – Focused on key parameters
 - Model field plate as a transistor and perform capacitance parameter extraction
 - RF modeling
 - Broadband S-Parameters and Y-Parameters
 - Large signal load pull and harmonic balance power sweeps
- The model is developed by using Keysight's IC-CAP and ADS software. Good fitting is achieved with industry measured data.

References

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